

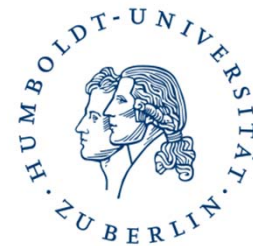
# Drahtlose Breitbandkommunikation - Kanalkapazität



innovations  
for high  
performance  

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microelectronics



**Humboldt-Universität zu Berlin, Institut für Informatik,  
IHP, Leibniz Institut für innovative Mikroelektronik, Frankfurt (Oder)**

Vorlesung Drahtlose Breitbandkommunikationssysteme

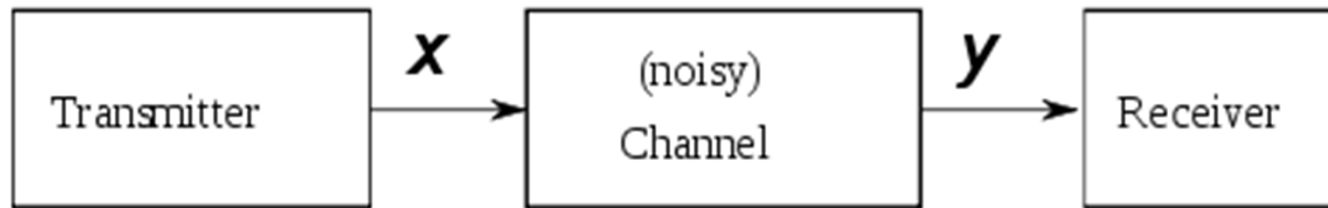
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# Channel Capacity: Shannon Theorem [Wikipedia]



- The **Shannon limit** or **Shannon capacity** of a communications channel is the theoretical maximum information transfer rate of the channel, for a particular noise level.
- Stated by Claude Shannon in 1948 (with sketchy proof)
- The first rigorous proof by Amiel Feinstein in 1954.
- The Shannon theorem states that given a noisy channel with channel capacity  $C$  and information transmitted at a rate  $R$ , then if  $R < C$  there exist codes that allow the probability of error at the receiver to be made arbitrarily small.
  - This means that, theoretically, it is possible to transmit information nearly without error at any rate below a limiting rate,  $C$ .

# Shannon-Hartley Theoreme (for AWGN)

$$C = B \log_2 \left( 1 + \frac{S}{N} \right)$$

- C*: Channel capacity (max. data rate) in bits per second;
- B*: Bandwidth of the channel in Hertz;
- S*: Total received signal power over the bandwidth (in case of a modulated signal, often denoted *C*, i.e. modulated carrier), measured in Watt or (Volt)<sup>2</sup>;
- N*: Total noise or interference power over the bandwidth, measured in Watt or (Volt)<sup>2</sup>; and
- S/N*: Signal-to-noise ratio (SNR) or the carrier-to-noise ratio (CNR) of the communication signal to the Gaussian noise interference expressed as a linear power ratio (not as logarithmic decibels).

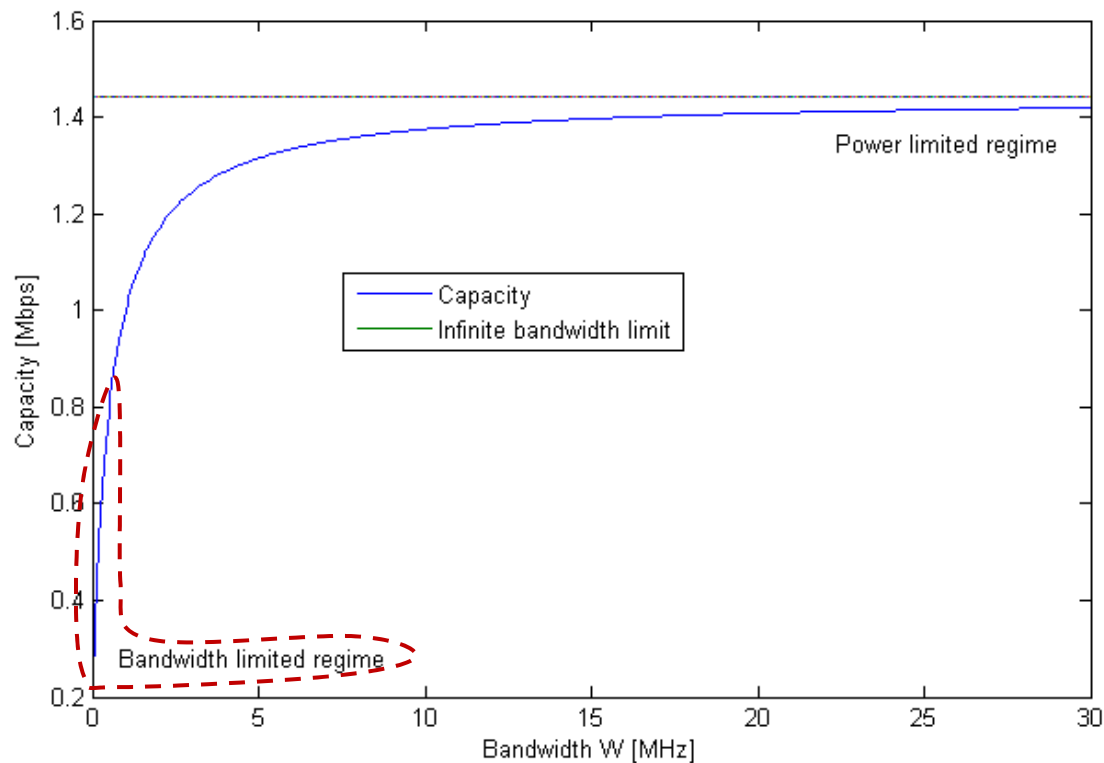
# AWGN Channel Capacity

$$C_{awgn} = W \log_2 \left( 1 + \frac{\bar{P}}{N_0 W} \right) \text{ [bits/s]}$$

- If the average received power is  $\bar{P}$  [W] and the noise power spectral density is  $N_0$  [W/Hz], the AWGN channel capacity is  $C_{awgn}$
- $W$  is the used bandwidth ( $\sim B$ )
- $\bar{P}/(N_0 * W)$  is the signal-to-noise-ratio (SNR) at the receiver

# Channel Capacity in Bandwidth-Limited Regime

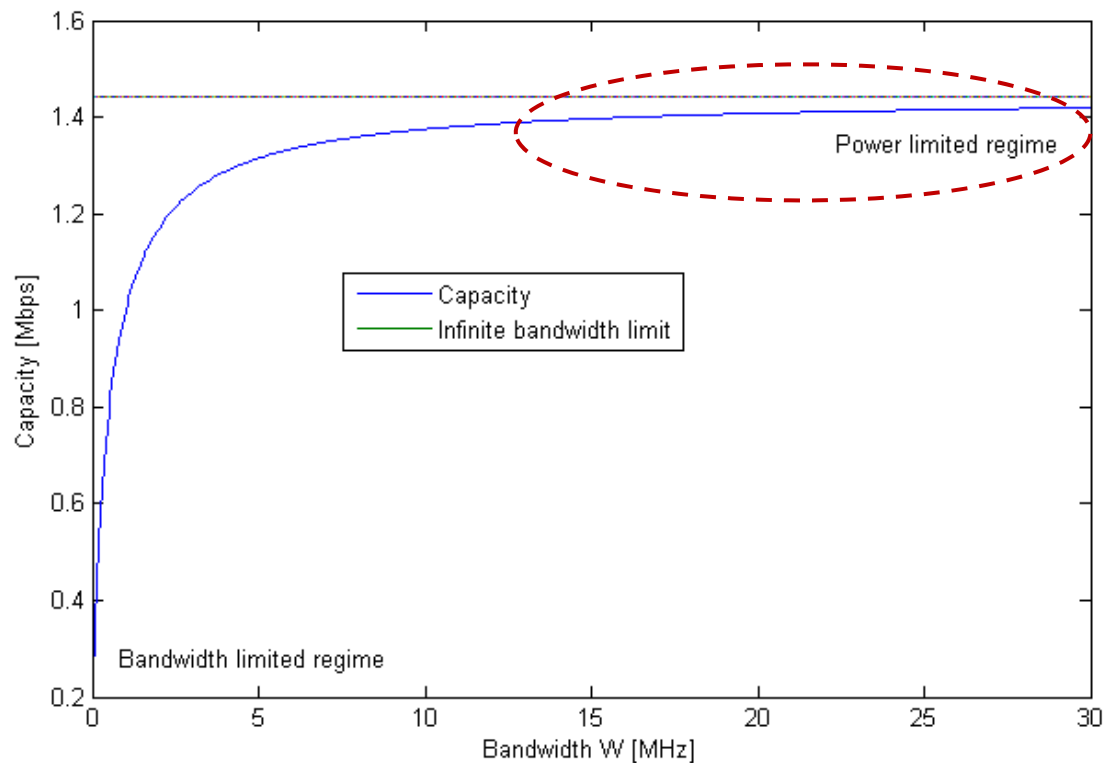
When the SNR is large (SNR  $\gg$  0 dB):  $C \approx W \log_2 \frac{\bar{P}}{N_0 W}$



- C is logarithmic in power and approximately linear in bandwidth  
-> *bandwidth-limited regime*

# Channel Capacity in Power-Limited Regime

When the SNR is small (SNR  $\ll$  0 dB):  
( $W \rightarrow \infty$ )  $C \approx \frac{\bar{P}}{N_0} \log_2 e$



- C is linear in power but insensitive to bandwidth  
-> *Power-limited regime*

# Approximations

For large or small and constant signal-to-noise ratios, the capacity formula can be approximated:

- If  $S/N \gg 1$ , then:  $C \approx 0.332 \cdot B \cdot \text{SNR (in dB)}$   
where:  $\text{SNR (in dB)} = 10 \log_{10} \frac{S}{N}$ .
- Similarly, if  $S/N \ll 1$ , then:  $C \approx 1.44 \cdot B \cdot \frac{S}{N}$ .

In this low-SNR approximation, capacity is independent of bandwidth if the noise is white, of spectral density  $N_0$  [W/Hz]. In this case the total noise power is  $B \cdot N_0$  and the channel capacity becomes:

$$C \approx 1.44 \cdot \frac{S}{N_0}$$

Source: [http://en.wikipedia.org/wiki/Shannon-Hartley\\_theorem](http://en.wikipedia.org/wiki/Shannon-Hartley_theorem)

## Frequency-dependent (colored noise) case

A generalization of the channel capacity equation for the case where the noise is not white (or that the S/N is not constant with frequency over the bandwidth) is obtained by treating the channel as many narrow, independent Gaussian channels in parallel:

$$C = \int_0^B \log_2 \left( 1 + \frac{S(f)}{N(f)} \right) df$$

$C$ : channel capacity in bits per second;

$B$ : bandwidth of the channel in Hz;

$S(f)$ : signal power spectrum

$N(f)$ : noise power spectrum

$f$ : frequency in Hz (within  $B$ )

Source: [http://en.wikipedia.org/wiki/Shannon-Hartley\\_theorem](http://en.wikipedia.org/wiki/Shannon-Hartley_theorem)

# Noise Power Spectral Density $N_o$

$$C_{awgn} = W \log_2 \left( 1 + \frac{\bar{P}}{N_o W} \right)$$

$$N_o = k * T \text{ [W/Hz]}$$

$k$ : Boltzmann constant [J/K] ( $= 1.38 \times 10^{-23}$  J/K)

$T$ : Temperature [K] ( $\sim 300$  K @ room temp.)

$$\begin{aligned} N_o &= k * T = 1.38 \times 10^{-23} \text{ J/K} * 300 \text{ K} \\ &= 4.14 * 10^{-21} \text{ J} \\ &= 4.14 * 10^{-21} \text{ Ws} \quad (\text{for room temperature}) \end{aligned}$$

## Example I (Telephone Modem)

- Example I:
  - Consider the operation of a modem on an ordinary telephone line. The SNR is usually about 1000 (30 dB). The bandwidth is 3.4 kHz.

Therefore:

$$\begin{aligned}C &= 3.4 \text{ kHz} * \log_2(1 + 1000) \text{ bit} \\ &= 3400 \text{ Hz} * 9.97 \text{ bit} \\ &= \underline{34 \text{ kbit/s}}\end{aligned}$$

*Source: Rahul Mangharam, Lecture ESE680: Wireless Sensor Networks*

## Example II (WLAN)

- Example II:
  - Consider the operation of a WLAN modem IEEE802.11a. The channel spacing is 20 MHz and the used bandwidth is approximately 16 MHz. The SNR for a good channel and low distance can be about 100 (= 20 dB).

Therefore:

$$\begin{aligned}C &= 16 \text{ MHz} * \log_2(1 + 100) \text{ bit} \\ &\approx 16 * 10^6 \text{ Hz} * 6.66 \text{ bit} \\ &= \underline{106.6 \text{ Mbit/s}}\end{aligned}$$

# Friis free-space path-loss equation

Received Power:

- $P_{\text{rcvd}}(d) = P_{\text{tx}} + G_t + G_r - \text{PL}$  [all in dB]

$P_{\text{tx}}$  = Transmit Power  
 $G_t$  = Gain of Transmit Antenna  
 $G_r$  = Gain of Receive Antenna  
PL = Path Loss

Path loss (PL) in free space:

$$\text{PL(a)} = (4 \cdot \pi \cdot d \cdot f / c)^2 \quad (\text{direct form})$$

$$\text{PL(log)} = 10 \cdot \log_{10}((4 \cdot \pi \cdot d \cdot f / c)^2) = 20 \cdot \log_{10}(f) + 20 \cdot \log_{10}(d) + 20 \cdot \log_{10}(4\pi/c)$$

(logarithmic form)

$d_0$

- d: Distance between Sender and Receiver
- f: Carrier Frequency
- c: Speed of light (~ 300 000 km/s)

# Friis Equation: Derivation @ board

- Friis Equation
- Link Budget Calculation