

Drahtlose Breitbandkommunikation – Link Budget



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Vorlesung Drahtlose Breitbandkommunikationssysteme

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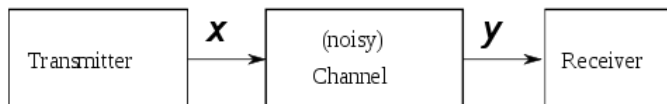
<http://www.informatik.hu-berlin.de/~grass/bbk>

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Drahtlose Breitbandkommunikation (BBK)

WS2013/14

Channel Capacity: Shannon Theorem [Wikipedia]



- The **Shannon limit** or **Shannon capacity** of a communications channel is the theoretical maximum information transfer rate of the channel, for a particular noise level.
- Stated by Claude Shannon in 1948 (with sketchy proof)
- The first rigorous proof by Amiel Feinstein in 1954.
- The Shannon theorem states that given a noisy channel with channel capacity C and information transmitted at a rate R , then if $R < C$ there exist codes that allow the probability of error at the receiver to be made arbitrarily small.
 - This means that, theoretically, it is possible to transmit information nearly without error at any rate below a limiting rate, C .

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Link Budget Calculation

Path loss (PL) in free space (Friis equation):

$$PL(a) = (4 \cdot \pi \cdot d \cdot f / c)^2 \quad (\text{direct form})$$

$$PL(\log) = 10 \cdot \log_{10}(4 \cdot \pi \cdot d \cdot f / c)^2 = 20 \cdot \log_{10}(f) + 20 \cdot \log_{10}(d) + 20 \cdot \log_{10}(4 \pi / c) \quad (\text{logarithmic form})$$

- d: Distance between Sender and Receiver
- f: Carrier Frequency
- c: Speed of light (~ 300 000 km/s)

Shannon channel capacity:

$$C = B \log_2 (1 + S/N)$$

d_0

Path Loss Exponent

Path loss (PL) in free space (Friis equation):

$$PL(a) = (4 \cdot \pi \cdot d \cdot f / c)^{2} \quad (\text{direct form})$$

- The additional losses caused by fading, blockage ... can be simply modeled by modifying the path loss exponent E_{PL}
- The path loss exponent $E_{PL} = 2$ for free space, line-of sight scenario
- For indoor and urban areas the path loss exponent can range from $E_{PL} = 2 \dots 4$
- In some lossy environment, the path loss exponent can reach even $E_{PL} = 6$
- In a specific tunnel scenario the path loss exponent can be $E_{PL} < 2$

Link Budget Calculation

- | | | |
|---|----------|-------|
| - TX-Power | P_{TX} | |
| - Friis equation (path loss) | L_{FS} | (=PL) |
| - Antenna gain TX | G_{TX} | |
| - Antenna gain RX | G_{RX} | |
| - RX Losses (cables, connectors, ...) | L_{RX} | |
| - Miscellaneous losses
(Implementation losses) | L_M | |
| | | |
| - RX-Noise Figure | NF | |
| - Shannon channel capacity | C | |

Received Power (all in dB):

$$P_{RX} \text{ (dB)} = P_{TX} + G_{TX} + G_{RX} - L_{FS} - L_{RX} - L_M$$

Required Signal/Noise at the Receiver

- During transmission and reception noise will be added to the signal
 - Thermal noise (Boltzmann noise)
 - Receiver Noise (Noise Figure)
 - Transmitter Noise caused by non linearities
 - etc.
- The effect for the received signal is that, in the constellation diagram, not a single point but a cloud of points around the expected point are received.
- Due to these noise and distortion impacts, the signal/noise (S/N) ratio to receive a signal correctly is at least around 6 dB (in the case of BPSK).
- The required SNR is dependent on the modulation and the transmission type
 - For 64 QAM, the SNR is at least 4 dB higher than for 16 QAM etc.

Receiver Sensitivity

The receiver sensitivity S_{\min} is the minimum input signal power required at the receiver input to provide **sufficient** SNR at receiver output to do data demodulation for a specific modulation scheme

- S_{\min} depends on:
 - Required SNR for desired modulation = SNR_{req}
 - Thermal noise floor at RX antenna = kT_oB
 - Noise Figure of the receiver = N_F

$$S_{\min} = \text{SNR}_{\text{req}} * kT_oB * N_F \quad (\text{all in absolute numbers})$$

$$S_{\min} \text{ (dB)} = \text{SNR}_{\text{req}} \text{ (dB)} + 10 * \log_{10}(kT_oB) + N_F \text{ (dB)}$$

- In some (generic) cases: $\text{SNR}_{\text{req}} = 0 \text{ dB}$

SNR <-> E_b/N_o

For the evaluation of the efficiency of a specific implementation, the parameter E_b/N_o is often used.

$$E_b/N_o = \text{SNR} * (B_T/R)$$

$$\text{SNR} = (E_b/N_o) * (R/B_T)$$

- SNR = Signal-to-noise ratio
- E_b = Energy required per bit of information
- N_o = thermal noise in 1Hz of bandwidth
- R = system data rate
- B_T = system bandwidth
- (R/B_T) = spectral efficiency

Source: Intersil, Tutorial Link budget Analysis 1998

Required Signal/Noise at the Receiver

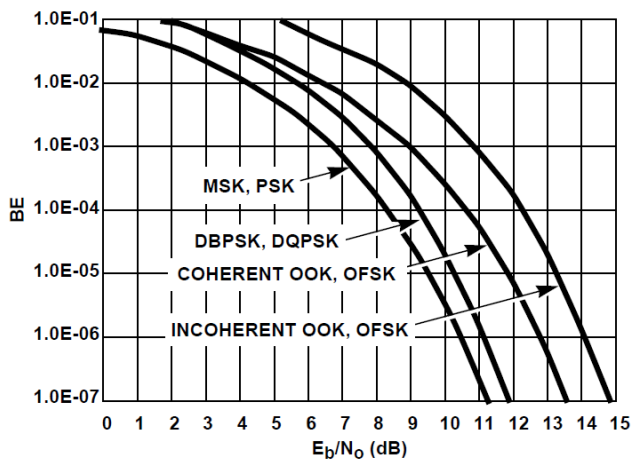


FIGURE 4. PROBABILITY OF BIT ERROR FOR COMMON MODULATION METHODS

Source: Intersil, Tutorial Link budget Analysis 1998

Link-Budget Calculator

- Excel Sheet

OFDM System Parameter Choice

Channel Parameters (T_{RMS} , Mobility, ...)

- Strong multipath propagation and high bit rate
 - OFDM (otherwise single carrier scheme sufficient)
- Delay spread of channel T_{RMS} determines :
 - guard time (cyclic prefix): $T_{\text{CP}} \sim 2 \dots 4 \times T_{\text{RMS}}$
 - $T_{\text{OFDM}} \sim > 5 \times T_{\text{CP}} \quad \rightarrow (T_{\text{sym}} = T_{\text{CP}} + T_{\text{FFT}})$
 - $T_{\text{FFT}} \sim > 4 \times T_{\text{CP}}$
- Number of subcarriers:
 - $N_{\text{sub}} = B \times T_{\text{FFT}}$
 - $N_{\text{sub}} = \text{target bitrate} / (\text{bitrate/subcarrier}) \quad (-> \text{modulation scheme})$
- Channel Coherence Time T_{C} determines:
 - Pilot scheme (time axis)
- Channel Coherence Bandwidth B_{C}
 - Pilot scheme (frequency axis)

OFDM System Parameter Choice II

Example Parameters for OFDM-parameter choice (@ board)

Channel properties / target parameters:

- Application scenario:
Large office; metal coated windows
- Delay spread of channel: $T_{\text{RMS}} = 200 \text{ ns}$
- Available bandwidth: $B = 20 \text{ MHz}$
- Target data rate: $R = 50 \text{ Mb/s}$
- Coherence time of channel: $T_{\text{C}} = 5 \mu\text{s}$
- Coherence bandwidth of channel: $B_{\text{C}} = 5 \text{ MHz}$